Zero-Drift, Bi-directional Current Sense Amplifier Description

The **BM181** series of zero-drift, bi-directional current sense amplifier can sense voltage drops across shunts at common-mode voltages from -0.3V to 36V, independent of the supply voltage. Three fixed gains are available: 50V/V, 100V/V and 200V/V. The low offset of the zero-drift architecture enables current sensing with maximum drops across the shunt as low as **10mV** full-scale.

BM181 devices operate from a single +2.7V to 30V power supply, with drawing a typical of 120uA of supply current. All versions are specified from –40°C +125°C, and offered in SC70-6 packages.

GAIN OPTIONS TABLE

	PRODUCT	GAIN	R3 and R4	R1 and R2	
	BM181	50	20kΩ	1ΜΩ	
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 $V_{OUT} = (I_{LOAD} \times R_{SHUNT})GAIN + V_{REF}$

高侧电流检测,可用于限流保护或恒流

Features

■ VOLTAGE OFFSET: ±100uV (MAX)

■ WIDE COMMON MODE VOLTAGE: -0.3V to +36V

■ SUPPLY VOLTAGE: 2.7V to +30V

ACCURACY and ZERO-DRIFT PERFORMANCE

♦ ±1% Gain Error (Max over temperature)

♦ 0.5µV/°C Offset Drift (Max)

◆ 10ppm/°C Gain Drift (Max)

■ THREE GAIN OPTIONS for VOLTAGE OUTPUT

◆ BM181: 50V/V

■ LOW SUPPLY CURRENT: 120uA (TYP)

■ Rail-to-Rail Output

■ PACKAGE: SC70-6

■ Industrial –40°C to 125°C Operation Range

■ ESD Rating: Robust 2KV – HBM, 2KV – CDM

Applications

CURRENT SENSING (High-Side/Low-Side)

BATTERY CHARGERS

■ POWER MANAGEMENT

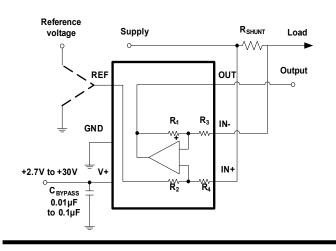
■ CELL PHONE CHARGER

■ ELECTRICAL CIGIRATE

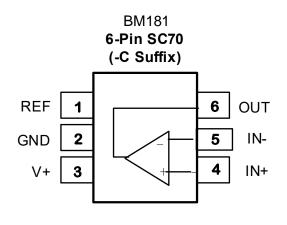
■ WIRELESS CHARGER

■ TELECOM EQUIPMENT

Application schematic



Pin Configuration



Order Information

Model Name	Order Number	Gain	Package	Transport Media, Quantity	Package Marking
BM181	BM181	50V/V	6-Pin SC70	Tape and Reel, 3,000	9A1

Absolute Maximum Ratings Note 1

Supply Voltage Note 2	42.0V	Current at Supply Pins	±60mA
Input Voltage	GND- 0.3 to 42V	Operating Temperature Range	40°C to 125°C
Input Current: +IN, -IN Note 3	±5mA	Maximum Junction Temperature	150°C
Output Current: OUT	±35mA	Storage Temperature Range	65°C to 150°C
		Lead Temperature (Soldering, 10 sec	;) 260°C

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The op amp supplies must be established simultaneously, with, or before, the application of any input signals.

Note 3: The inputs are protected by ESD protection diodes to each power supply. If the input extends more than 500mV beyond the power supply, the input current should be limited to less than 10mA.

ESD, Electrostatic Discharge Protection

Symbol	Parameter	Condition	Minimum Level	Unit
НВМ	Human Body Model ESD	ANSI/ESDA/JEDEC JS-001	±2	kV
CDM	Charged Device Model ESD	ANSI/ESDA/JEDEC JS-002	±2	kV

Thermal Resistance

Package Type	θ _{ЈА}	θυς	Unit
6-Pin SC70	227	80	°C/W

Electrical Characteristics

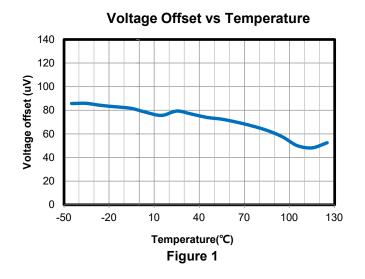
The specifications are at TA = 25°C, VSENSE = VIN+ – VIN–, VS = 5 V, VIN+ = 12V, and VREF = VS / 2, unless otherwise noted

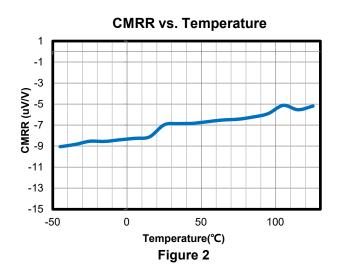
Vos TC Input Offset Voltage Drift	Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
Vos TC	INPUT							
V _{CM} Common-mode Input Range	Vos	Input Offset Voltage	VSENSE = 0 mV		±10	±100	uV	
CMRR Common Mode Rejection Ratio VIN+ = 5-26 V, VSENSE = 0 mV, -40°C to 125°C 95 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120 120	V _{os} TC	Input Offset Voltage Drift	VSENSE = 0 mV, -40°C to 125°C		0.1	0.5	μV/°C	
Input Bias Current VSENSE = 0 mV 0.4 0.4 0.5	V _{CM}	Common-mode Input Range	-40°C to 125°C	-0.3		36	V	
Input Offset Current VSENSE = 0 mV	CMRR	Common Mode Rejection Ratio	VIN+ = 5~26 V, VSENSE = 0 mV, -40°C to 125°C	95	120		dB	
PSRR Power Supply Rejection Ratio Vs = +2.7~18V, VIN+ = +18V, VSENSE = 0 mV ±1 u u v NOISE RTI Note 4 w NOISE RTI Note 4 w w w NOISE RTI Note 4 w w w w w w w w w	I _B	Input Bias Current	VSENSE = 0 mV		35		uA	
NOISE RTI Note 4	I _{os}	Input Offset Current	VSENSE = 0 mV		0.4		uA	
en Input Voltage Noise Density f = 1kHz 30 nV OUTPUT BM181 50 N GE Gain 50 N GE Gain Error VSENSE = -5~5mV, -40°C to 125°C ±0.1% ±1% GE TC Gain Error Vs Temperature -40°C to 125°C 3 10 p C _{LOAD} Maxim capacitive load No oscillation 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	PSRR	Power Supply Rejection Ratio	Vs = +2.7~18V, VIN+ = +18V, VSENSE = 0 mV		±1		uV/V	
OUTPUT G Gain BM181 50 V GE Gain Error VSENSE = -5-5mV, -40°C to 125°C ±0.1% ±1% GETC Gain Error Vs Temperature -40°C to 125°C 3 10 p CLOAD Maxim capacitive load No oscillation 1 1 1 1 VOH Output Swing from Supply Rail RLOAD = 10kΩ to REF, -40°C to 125°C 0.02 0.05 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	NOISE RTI No	ote 4						
BM181 50 No.	e _n	Input Voltage Noise Density	f = 1kHz		30		nV/√Hz	
Gain Gain	ОИТРИТ							
GE Gain Error VSENSE = -5~5mV, -40°C to 125°C ±0.1% ±1% GE TC Gain Error Vs Temperature -40°C to 125°C 3 10 p C _{LOAD} Maxim capacitive load No oscillation 1 V _{OH} Output Swing from Supply Rail R _{LOAD} = 10kΩ to REF, -40°C to 125°C 0.02 0.05 V _{OL} Output Swing from Supply Rail R _{LOAD} = 10kΩ to REF, -40°C to 125°C 0.01 0.05 FREQUENCY RESPONSE CLOAD = 10pF, BM181 48 k BW Bandwidth SR Slew Rate 0.6 V POWER SUPPLY V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 p			BM181		50		V/V	
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$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$								
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	GE	Gain Error	VSENSE = -5~5mV, -40°C to 125°C		±0.1%	±1%		
V _{OH} Output Swing from Supply Rail $R_{LOAD} = 10k\Omega$ to REF, -40°C to 125°C 0.02 0.05 FREQUENCY RESPONSE BW Bandwidth CLOAD = 10pF, BM181 48 k SR Slew Rate 0.6 V POWER SUPPLY V+ Supply Voltage 2.7 30 Iq Quiescent Current VSENSE = 0 mV 120 150 Ip	GE TC	Gain Error Vs Temperature	-40°C to 125°C		3	10	ppm	
V_{OL} Output Swing from Supply Rail $R_{LOAD} = 10$ kΩ to REF, -40°C to 125°C 0.01 0.05 FREQUENCY RESPONSE CLOAD = 10pF, BM181 48 k BW Bandwidth 0.6 V POWER SUPPLY V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 p	C _{LOAD}	Maxim capacitive load	No oscillation		1		nF	
FREQUENCY RESPONSE BW Bandwidth CLOAD = 10pF, BM181 48 k SR Slew Rate 0.6 V POWER SUPPLY V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 I _I	V _{OH}	Output Swing from Supply Rail	R_{LOAD} = 10kΩ to REF, -40°C to 125°C		0.02	0.05	V	
CLOAD = 10pF, BM181	V _{OL}	Output Swing from Supply Rail	R_{LOAD} = 10kΩ to REF, -40°C to 125°C		0.01	0.05	V	
BW Bandwidth 0.6 V SR Slew Rate 0.6 V POWER SUPPLY V+ Supply Voltage 2.7 30 Iq Quiescent Current VSENSE = 0 mV 120 150 Ip	FREQUENCY	RESPONSE						
SR Slew Rate 0.6 V POWER SUPPLY V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 1			CLOAD = 10pF, BM181		48		kHz	
POWER SUPPLY V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 1	BW	Bandwidth						
POWER SUPPLY V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 1								
V+ Supply Voltage 2.7 30 I _Q Quiescent Current VSENSE = 0 mV 120 150 µ	SR	Slew Rate			0.6		V/µs	
I _Q Quiescent Current VSENSE = 0 mV 120 150 µ	POWER SUPPLY							
	V+	Supply Voltage		2.7		30	V	
TEMPEDATURE DANCE	ΙQ	Quiescent Current	VSENSE = 0 mV		120	150	μA	
TEMPERATURE RANGE	TEMPERATU	IRE RANGE						
Specified range -40 125		Specified range		-40		125	°C	
Operating range -55 150		Operating range		-55		150	°C	

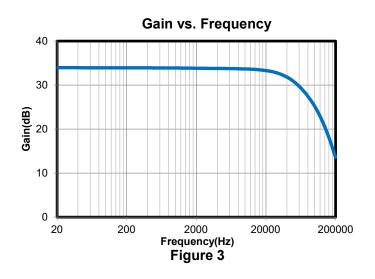
Note 4: RTI = referred to input

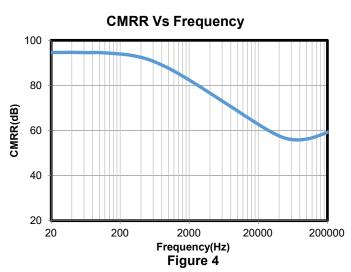
Typical Performance Characteristics

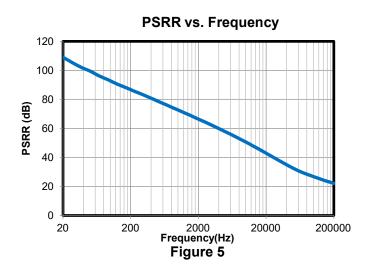
The BM181 is used for characteristics at TA = 25°C, VS = 5V, VIN+ =12V, and VREF=VS/2, unless otherwise noted











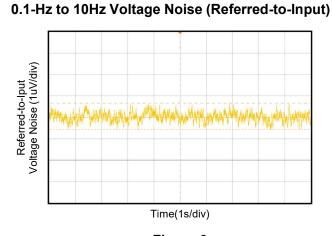
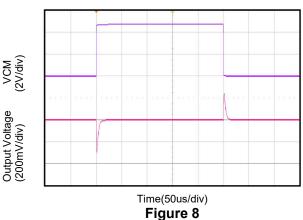


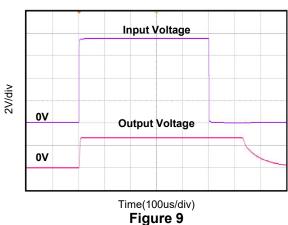
Figure 6

Typical Performance Characteristics

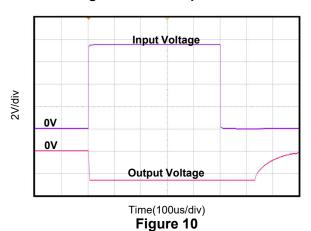
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Common-Mode Voltage Transient Response

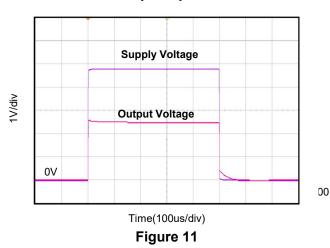




Inverting Differential Input Overload



Start-up Response



Brownout Recovery

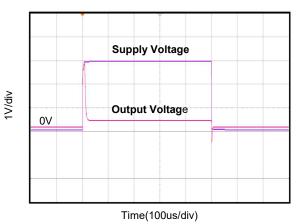
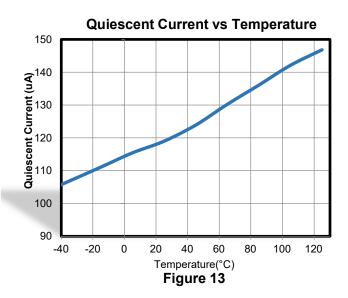


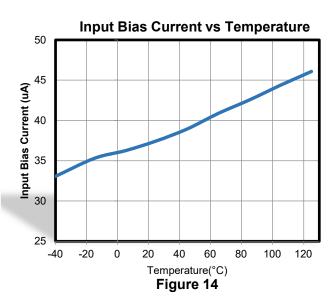
Figure 12

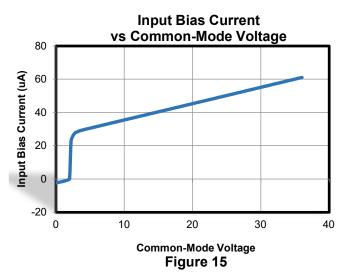
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Typical Performance Characteristics

The BM181 is used for characteristics at TA = 25°C, VS = 5V, VIN+ =12V, and VREF=VS/2, unless otherwise noted







Pin Functions

IN-: Inverting Input of the Amplifier.

IN+: Non-Inverting Input of Amplifier.

OUT: Amplifier Output. The voltage range extends to within mV

of each supply rail.

REF: Reference voltage

V+: Positive Power Supply. Typically, the voltage is from 2.7V to 30V. A bypass capacitor of 0.1µF as close to the part as possible should be used between power supply pin and ground pin.

GND: Negative Power Supply.

Operation Overview

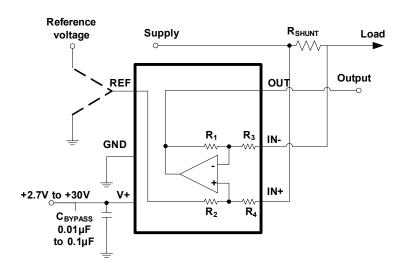
The BM181 family is 36V common-mode, zero-drift topology, current-sensing amplifiers that can be used in both low-side and high-side configurations. These specially-designed, current-sensing amplifiers are able to accurately measure voltages developed across current-sensing resistors on common-mode voltages that far exceed the supply voltage powering the device. Current can be

measured on input voltage rails as high as 36 V while the device can be powered from supply voltages as low as 2.7 V.

The zero-drift topology enables high-precision measurements with maximum input offset voltages as low as $100\mu V$ with a maximum temperature contribution of $0.5 \mu V/^{\circ}C$ over the full temperature range of $-40^{\circ}C$ to $125^{\circ}C$.

Applications Information

Application schematic



Above figure shows the basic connections of the BM181. The input pins, IN+ and IN-, should be connected as closely as possible to the shunt resistor to minimize any resistance in series with the shunt resistor.

Power-supply bypass capacitors are required for stability. Applications with noisy or high-impedance power supplies may require additional decoupling capacitors to reject power-supply noise. Connect bypass capacitors close to the device pins.

Selecting RSHUNT

The zero-drift offset performance of the BM181 offers several benefits. Most often, the primary advantage of the low offset characteristic enables lower full-scale drops across the shunt. For example, nonzero-drift current shunt monitors typically require a full-scale range of 100 mV.

The BM181 family gives equivalent accuracy at a full-scale range on the order of 10 mV. This accuracy reduces shunt dissipation by an order of magnitude with many additional benefits.

Alternatively, there are applications that must measure current over a wide dynamic range that can take advantage of the low offset on the low end of the measurement. Most often, these applications can use the lower gains of the BM181 to accommodate larger shunt drops on the upper end of the scale. For instance, an BM181 operating on a 3.3-V supply could easily handle a full-scale shunt drop of 60 mV, with only 100uV of offset.

REF Input Impedance Effects

As with any difference amplifier, the BM181 family common-mode rejection ratio is affected by any impedance present at the REF input. This concern is not a problem when the REF pin is connected directly to most references or power supplies. When using resistive dividers from the power supply or a reference voltage, the REF pin should be buffered by an op amp.

Power Supply Recommendation

The input circuitry of the BM181 can accurately measure beyond its power-supply voltage, V+. For example, the V+ power supply

can be 5 V, whereas the load power-supply voltage can be as high as 30 V. However, the output voltage range of the OUT pin is limited by the voltages on the power-supply pin. Note also that the BM181 can withstand the full input signal range up to 36 V at the input pins, regardless of whether the device has power applied or not.

Proper Board Layout

To ensure optimum performance at the PCB level, care must be taken in the design of the board layout. To avoid leakage currents, the surface of the board should be kept clean and free of moisture. Coating the surface creates a barrier to moisture accumulation and helps reduce parasitic resistance on the board.

Keeping supply traces short and properly bypassing the power supplies minimizes power supply disturbances due to output current variation, such as when driving an ac signal into a heavy load. Bypass capacitors should be connected as closely as possible to the device supply pins. Stray capacitances are a concern at the outputs and the inputs of the amplifier. It is recommended that signal traces be kept at least 5mm from supply lines to minimize coupling.

A variation in temperature across the PCB can cause a mismatch in the Seebeck voltages at solder joints and other points where dissimilar metals are in contact, resulting in thermal voltage errors. To minimize these thermocouple effects, orient resistors so heat sources warm both ends equally. Input signal paths should contain matching numbers and types of components, where possible to match the number and type of thermocouple junctions. For example, dummy components such as zero value resistors can be used to match real resistors in the opposite input path. Matching components should be located in close proximity and should be oriented in the same manner. Ensure leads are of equal length so that thermal conduction is in equilibrium. Keep heat sources on the PCB as far away from amplifier input circuitry as is practical.

The use of a ground plane is highly recommended. A ground plane reduces EMI noise and also helps to maintain a constant temperature across the circuit board.

Package Outline Dimensions

SC70-6 /SOT-363

